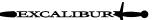
PT4560 Series—48V

30 Watt 48V Input Isolated DC-DC Converter



Power Trends Products from Texas Instruments

SLTS097

(Revised 6/30/2000)



The PT4560 is a new series of 30 Watt, isolated DC to DC converters housed in a unique 19-pin, thermally efficient package. The 36 to 75V input range allows easy integration into many distributed power applications which utilize 48V bus architectures.

The PT4560 series is available with output voltages from 1.8V to 15V. The output voltage is adjustable from 90 to 110% of nominal with the addition of an external resistor. Other features include an inhibit function and differential remote sense to compensate for any voltage drop between the converter and the load. The PT4560 includes built in current limit, short circuit protection and over-temperature shutdown.

A 330µF output capacitor is required for proper operation.

Standard Application

Characteristics

Output Current

(T_a = 25°C unless noted)

On/Off Standby Current

Output Voltage Tolerance

Short Circuit Current

Input Voltage Range

Line Regulation

Load Regulation

Vo Ripple/Noise

Efficiency

Transient Response

Switching Frequency

Maximum Operating Temperature Range

Storage Temperature

Mechanical Vibration

Input/Output Isolation Capacitance

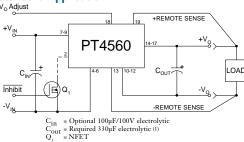
Mechanical Shock

Weight

Resistance

Flammability

Inhibit (pin 2)



Symbols

Iin standby

 I_{o}

I.

Vin

 $\Delta V_{\rm o}$

Regline

Regload

Vn

t_{tr}

η

 f_{o}

Ta

T,

Pin-Out Information

Pin Function	Pin Function
1 Do Not Use	10 -V _o
2 Inhibit ⁽⁴⁾	11 -V _o
3 Do Not Use	12 -V _o
4 -V _{in}	13 -Remote Sense
5 -V _{in}	14 +V _o
6 -V _{in}	15 +V _o
7 +V _{in}	16 +V _o
8 +Vin	17 +V _o
9 +Vin	18 V _o Adjust ⁽⁴⁾
	19 +Remote Sense

PT4560 SERIES

Max

8.0

6.0 2.5

16

75.0

±2.0

±1.0

±1.0

2.0 75

200

_

900

600

+85 (2)

+125

Тур

8

2xIor

48.0

±1.0

±0.5

±0.5

1.0

50

100

3.0

80

84 85

750 500

500

10

40

1200

Units

A

A A

mA

А

V

 $%V_{0}$

%V

%Vo

%Vo $m\underline{V}_{pp}$

μSec

ν%

%

%

kHz

°C

°C

G's

G's

grams

VDC

pF MΩ

Min

0.25

0.25

36.0

_

600

400

-40

1500

10

2.5

-40 (1)

0.1

 $V_0 = 3.3V$ $V_0 = 5V$

V₀<10V V₀≥10V

Features

- Input Voltage Range: 36V to 75V
- 1500 VDC Isolation
- V_o Inhibit
- V_oAdjust
- Differential Remote Sense
- Current Limit
- Short-Circuit Protection
- Over-Temperature Shutdown
- Undervoltage Lockout
- ExcaliburTM Package
- UL1950 recognized
- CSA 22.2 950 certified
- EN60950 Approved
- VDE Licensed
- Within FCC Class A Radiated Limits

Ordering Information

PT4561 □ = 3.3V/8A
PT4562 □ = 5.0V/6A
PT4563 □ = 12.0V/2.5A
PT4564 □ = 15.0V/2A
PT4565 □ = 2.0V/8A
PT4566 □ = 2.5V/8A
PT4567 □ = 1.8V/8A
PT4568 □ = 5.2V/6A

PT Series Suffix (PT1234X) Case/Pin

Configuration

Vertical Through-Hole Horizontal Through-Hole А С Horizontal Surface Mount

(For dimensions and PC board layout, see Package Styles 1400 and 1410.)

Note: This product is the subject of one or more patents. Other patents pending.

On (3) Off 15 0.8 VDC VDC NOTES: (1) At temperatures below 0°C the PT4560 requires capacitors with temperature stable dielectrics, such as tantalum or Oscon. (2) See Safe Operating Area curves, or cantact the factory for the appropriate derating.

(3) If pin 2 is left open, the PT4560 will operate when input power is applied.

Conditions

 $V_{in} = 48V$

Over I_o Range

Over Vin range

 $V_{in} = 48V$, Pin 1 = - V_{in}

Over V_{in} Range T_A= -40°C to +85°C

Over Vin range @ max Io

10% to 100% of Io max

 V_{in} =48V, I_o = $I_o max$, $V_o \ge 5V$ V_{in} =48V, I_o = $I_o max$, $V_o < 5V$

50% load change, 1A/µSec

 V_o over/undershoot, $V_{o\geq}5V$

Per Mil-Std-883D, method 2002.3,

1mS, half-sine, mounted to a fixture

Per Mil-Std-883D, method 2007.2

20-2000Hz, soldered in a PC board

Materials meet UL 94V-0

Referenced to -Vin

Over Vin and Io

V_{in} = 48V @ max I_o Airflow = 200 LFM

(4) See Application Notes at end of section.

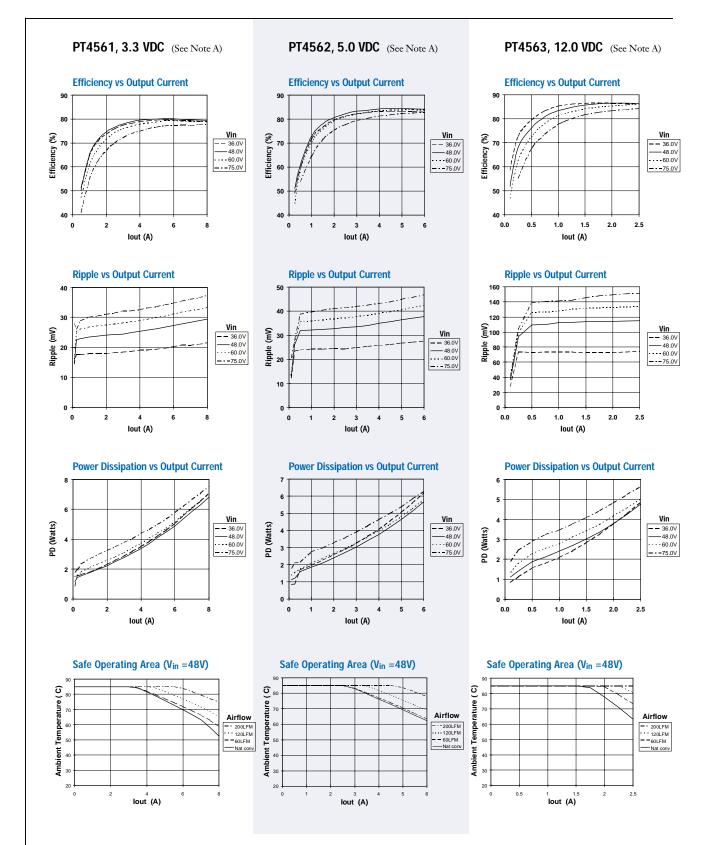




Preliminary Specifications

Typical Characteristics

30 Watt 48V Input Isolated DC-DC Converter



Note A: All data listed in the above graphs has been developed from actual products tested at 25°C. This data is considered typical data for the DC-DC Converter. Note B: SOA curves represent operating conditions at which internal components are at or below manufacturer's maximum rated operating temperature.

PT3320/3340/4560/4580 Series

Adjusting the Output Voltage of Power Trends' 30W Isolated DC-DC Converter Series

The factory pre-set output voltage of Power Trends' 30W series of isolated DC-DC converters may be adjusted within a nominal $\pm 10\%$ range. This is accomplished with the addition of a single external resistor. For the input voltage range specified in the data sheet, Table 1 gives the allowable adjustment range for each model as V_o (min) and V_o (max).

Adjust Up: An increase in the output voltage is obtained by adding a resistor, R2 between pin 18 (V_o adjust), and pin 13 (-Remote Sense). See note 4.

Adjust Down: Add a resistor (R1), between pin 18 (V_o adjust) and pin 19 (+Remote Sense).

Refer to Figure 1 and Table 2 for both the placement and value of the required resistor, (R1) or R2.

Notes:

- 1. Use only a single 1% resistor in either the (R1) or R2 location. Place the resistor as close to the ISR as possible.
- 2. Never connect capacitors to $V_{_{\rm o}}$ adjust. Any capacitance added to the $V_{_{\rm o}}$ adjust control pin will affect the stability of the ISR.
- If the remote sense pins are not being used, the resistors (R1) and R2 can be connected to +V_{out} or -V_{out} respectively.

Table 1

DC-DC CONVERTER ADJUSTMENT RANGE AND FORMULA PARAMETERS						
Series Pt #						

4. The adjusted output voltage, V_a effectively sets the voltage across pins 13 and 19 (±Remote Sense). When using the remote sense pins, V_{out} (measured directly across pins 10–12, and 14–17) can be significantly higher than V_a, and may exceed V_o (max). If V_a is adjusted upward of V_o(max), the the minimum input voltage is increased by the same percentage as V_{out} exceeds V_o(max).

The values of (R1) [adjust down], and R2 [adjust up], can also be calculated using the following formulas.

$$(\mathbf{R1}) = \frac{\mathbf{K}_{o} (\mathbf{V}_{a} - \mathbf{V}_{r})}{\mathbf{V}_{r} (\mathbf{V}_{o} - \mathbf{V}_{a})} - \mathbf{R}_{s} \qquad \mathbf{k} \Omega$$

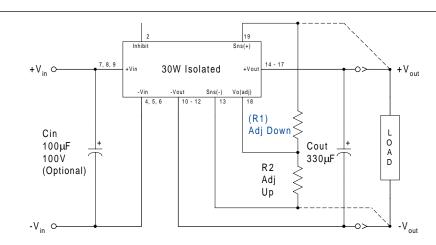
$$R2 \quad = \quad \frac{K_o}{(V_a - V_o)} \quad - R_s \qquad k\Omega$$

Where V_o = Original output voltage

- V_a = Adjusted output voltage
- V_r = Reference voltage (Table 1)
- K_o = Multiplier constant (Table 1)
- R_s = Series resistance (Table 1)

Series Pt #							
AL Case:							
24V Bus				PT3341	PT3342	PT3343	PT3344
48V Bus	PT3327	PT3325	PT3326	PT3321	PT3322	PT3323	PT3324
CU Case:							
24V Bus	PT4585			PT4581	PT4582	PT4583	PT4584
48V Bus	PT4567	PT4565	PT4566	PT4561	PT4562	PT4563	PT4564
V _o (nom)	1.8V	2.0V	2.5V	3.3V	5.0V	12.0V	15.0V
Vo(min)	1.62V	1.8V	2.25V	2.95V	4.5V	10.8V	13.5V
Vo(max)	1.98V	2.2V	2.75V	3.65V	5.5V	13.2V	16.5V
Vr	1.225V	1.225V	1.225V	1.225V	1.225V	2.5V	2.5V
K ₀ (V·k Ω)	69.58	62.47	42.33	68.89	68.71	135.9	137.5
Rs (kΩ)	80.6	150.0	121.0	150.0	121.0	90.9	80.6

Figure 1





PT3320/3340/4560/4580 Series

	ERTER ADJUSTN	VENT RESISTOR	R VALUES						
eries Pt #									
AL Case			DT2241	DT2242		DT2242		DT2244	
24V Bus 48V Bus	PT3327	PT3325	PT3341 PT3326	PT3342 PT3321		PT3343 PT3322		PT3344 PT3323	PT3324
CU Case	F13327	F13323	F13320	F13321		FIJJZZ		F13323	F13324
24V Bus	PT4585			PT4581		PT4582		PT4583	PT4584
48V Bus	PT4567	PT4565	PT4566	PT4561		PT4562		PT4563	PT4564
Current	8Adc	8Adc	8Adc	8Adc		6adc	_	2.5Adc	2.0Adc
I₀(nom)	1.8V	2.0Vdc	2.5Vdc	3.3Vdc		5.0Vdc		12.0Vdc	15.0Vdc
l _a (req'd)					V _a (req'd)		V _a (req'd)		
1.65	(39.9)kΩ				4.5	(246.0)kΩ	10.8	(285.0)kΩ	
1.7	(149.0)kΩ				4.55	(293.0)kΩ	11.0	(371.0)kΩ	
1.75	(475)kΩ				4.6	(352.0)kΩ	11.2	(500.0)kΩ	
1.8					4.65	(428.0)kΩ	11.4	(715.0)kΩ	
1.85	1270.0kΩ	(62.5)kΩ			4.7	(529.0)kΩ	11.6	(1150.0)kΩ	
1.9	575.0kΩ	(194.0)kΩ			4.75	(670.0)kΩ	11.8		
1.95	343.0kΩ	(589.0)kΩ			4.8	(882.0)kΩ	12.0		
2.0					4.85	(1230.0)kΩ	12.2	588.0kΩ	
2.05		1100.0kΩ			4.9	(1940.0)kΩ	12.4	249.0kΩ	
2.1		475.0kΩ			4.95		12.6	136.0kΩ	
2.15		266.0kΩ			5.0		12.8	78.9kΩ	
2.2		162.0kΩ			5.05		13.0	45.0kΩ	
2.25			(20.7)kΩ		5.1	566.0kΩ	13.2	22.3kΩ	
2.3			(64.7.0)kΩ		5.15	337.0kΩ			
2.35			(138.0)kΩ		5.2	223.0kΩ	13.5		(323.0)kΩ
2.4			(285.0)kΩ		5.25	154.0kΩ	13.6		(355.0)kΩ
2.45			(726.0)kΩ		5.3	108.0kΩ	13.8		(437.0)kΩ
2.5					5.35	75.3kΩ	14.0		(522.0)kΩ
2.55			726.0kΩ		5.4	50.8kΩ	14.2		(724.0)kΩ
2.6			302.0kΩ		5.45	31.7kΩ	14.4		(1010.0)kΩ
2.65			161.0kΩ		5.5	16.4kΩ	14.6		(1580.0)kΩ
2.7			90.6kΩ				14.8		
2.75			48.3kΩ				15.0		
2.95				(127.0)kΩ			15.2		607.0kΩ
3.0				(183.0)kΩ			15.4		263.0kΩ
3.05				(261.0)kΩ			15.6		149.0kΩ
3.1				(377.0)kΩ			15.8		91.3kΩ
3.15				(572.0)kΩ			16.0		56.9kΩ
3.2				(961.0)kΩ			16.5		11.1kΩ
3.25				(2130.0)kΩ	_				
3.3				<u></u>					
3.35				1230.0kΩ					
3.4				539.0kΩ					
3.45				309.0kΩ					
3.5				194.0kΩ					
3.55				126.0kΩ					
3.6				79.6kΩ					
3.65				46.8kΩ					

R1 = (Blue) R2

V Texas Instruments

R2 = Black

For technical support and more information, see inside back cover or visit www.ti.com/powertrends

PT3320/3340/4560/4580 Series

Using the Inhibit Function on the Power Trends' 30W Isolated DC-DC Converter Series

For applications requiring output voltage on/off control, the Power Trends' 30W isolated series of DC-DC converters incorporate an inhibit function. This function may be used in applications that require battery conservation, power-up/shutdown sequencing, and/or to co-ordinate the power-up of the regulator for active in-rush current control. (See the related application note, AN21).

The inhibit function is provided by the *Inhibit* control, pin2. If pin 2 is left open-circuit, the converter provides a regulated output whenever a valid source voltage⁴ is applied between $+V_{in}$ (pins 7-9), and $-V_{in}$ (pins 4-6). Applying a low-level ground signal² to pin 2 will disable the regulator output. The inhibit control is also compatible with some logic families. Table 1 provides details of the logic threshold requirements for the inhibit input. Figure 1 shows how either a discrete MOSFET (Q₁) or a logic gate (U1_a)⁵, may be referenced to the negative input voltage rail and used with the inhibit control.

Table 1 Inhibit Control Thresholds ²

Parameter	min	max	
Enable (VIH)	2.5V	15.0V	(or Open Circuit) 5
Disable (VIL)	-0.3V	0.8V	

Notes:

- The inhibit control logic is similar for all Power Trends' modules, but the flexibility and threshold tolerances will be different. For specific information on this function for other converters, consult the applicable application note.
- 2. The inhibit control pin uses $-V_{in}$ (pins 4-6) as its ground reference. All voltages specified are with respect to $-V_{in}$.
- The inhibit control internal circuitry comprises of a high impedance 10μA current source. The open-circuit voltage may be as high as 8.3Vdc.
- 4. These converters incorporate an "Under Voltage Lockout" (UVLO) function. This function automatically inhibits the converter output until there is sufficient input voltage for the converter to produce a regulated output. Table 2 gives the applicable UVLO thresholds.

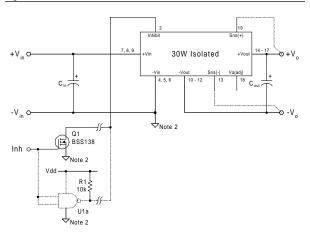
Table 2 UVLO Thresholds ^{2/}					
Series	UVLO Threshold	V _{in} Range			
PT3320/4560	33 ±2.0V	36 - 75 V			
PT3340/4580	$15.5 \pm 1.5 \text{V}$	18 60V			

- 5. The inhibit pin may be controlled using devices with either an open-collector or differential output. With a bipolar transistor or MOSFET, select a low-leakage part (<1 μ A). A pull-up resistor is not necessary. If a logic gate is used a pull-up resistor may be required to the logic supply voltage, V_{dd}. This is to ensure that the gate output exceeds V_{IH}(min) (see Table 1). <u>Do not</u> use a pull-up resistor to the +V_{in} input, or drive the inhibit pin above V_{IH}(max).
- 6. When the converter output is disabled the current drawn

from the input supply is typically reduced to 8mA (16mA maximum).

7. Keep the on/off transition to less than 1ms. This prevents erratic operation of the ISR, whereby the output voltage may drift un-regulated between 0V and the rated output during power-up.

Figure 1



Turn-On Time: When the Inhibit pin is left open-circuit, the output of the regulator is automatically enabled when a valid input voltage⁴ is applied to the input power pins. The converter typically produces a fully regulated output voltage within 30-ms of the application of power, or the release of the inhibit pin with input power applied. The actual turn-on time will vary with the input voltage, output load, and the total amount of capacitance connected to the output Using the circuit of Figure 1, Figure 2 shows the typical output voltage and input current waveforms of a PT3322/PT4562 after Q1 is turned off at time t = 0s. The waveform was measured with a 48Vdc input voltage, and 1-ohm resistive load.



